

SRV02-SeriesRotary Experiment # 2

Speed Control



Student Handout



SRV02-SeriesRotary Experiment # 2

Speed Control Student Handout

1. Objectives

The objective in this experiment is to design a lead controller that will regulate the speed of the output shaft.

At the end of this session, you should know the following:

- How to mathematically model the servo plant from first principles.
- To acquire an open-loop Bode plot of the system.
- To design and simulate a lead controller to meet the required specifications (using phase margin design).
- To implement your controller and evaluate its performance.

2. System Requirements

To complete this lab, the following hardware is required:

- [1] Quanser UPM 2405/1503 Power Module or equivalent.
- [1] Quanser MultiQ PCI / MQ3 or equivalent.
- [1] Quanser SRV02-T servo plant.
- [1] PC equipped with the required software as stated in the WinCon user manual.
- The suggested configuration for this experiment is the SRV02-T in the Low-Gear configuration with a UPM 1503 power module and a gain cable of 1.
- It is assumed that the student has successfully completed Experiment #0 of the SRV02 and is familiar in using WinCon to control the plant through Simulink.
- It is also assumed that all the sensors and actuators are connected as per dictated in the Experiment #0 as well as the SRV02 User's Manual.

Page # 2 Revision: 01

3. Mathematical Model

This section of the lab should be read over and completely understood before attending the lab. It is encouraged for the student to work through the derivations as well as to get a thorough understanding of the underlying mechanics. For a complete listing of the symbols used in this derivation as well as the model, refer to **Appendix A - SRV02 Nomenclature** at the back of this handout.

We shall begin by examining the electrical component of the motor first. In **Figure 1**, you see the electrical schematic of the armature circuit.

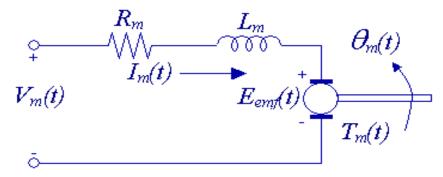


Figure 1 - Armature circuit in the time-domain

Using Kirchhoff's voltage law, we obtain the following equation:

$$V_m - R_m I_m - L_m \frac{dI_m}{dt} - E_{emf} = 0$$
 [3.1]

Since $L_m \ll R_m$, we can disregard the motor inductance leaving us with:

$$I_m = \frac{V_m - E_{emf}}{R_m}$$
 [3.2]

We know that the back emf created by the motor is proportional to the motor shaft velocity ω_m such that:

$$I_m = \frac{V_m - K_m \omega_m}{R_m}$$
 [3.3]

We now shift over to the mechanical aspect of the motor and begin by applying Newton's 2nd law of motion to the motor shaft:

$$J_m \dot{\omega}_m = T_m - \frac{T_l}{\eta_g K_g}$$
 [3.4]

Page # 3 Revision: 01

Where $\frac{T_l}{\eta_\sigma K_a}$ is the load torque seen thru the gears. And η_g is the efficiency of the gearbox.

We now apply the 2nd law of motion at the load of the motor:

$$J_l \dot{\omega}_l = T_l - B_{eq} \omega_l \tag{3.5}$$

Where \mathbf{B}_{eq} is the viscous damping coefficient as seen at the output.

Substituting [3.4] into [3.5], we are left with:

$$J_l \dot{\omega}_l = \eta_a K_a T_m - \eta_a K_a J_m \dot{\omega}_m - B_{ea} \omega_l$$
 [3.6]

We know that $\omega_m = K_g \omega_l$ and $T_m = \eta_m K_t I_m$ (where η_m is the motor efficiency), we can re-write [3.6] as:

$$J_l \dot{\omega}_l + \eta_a K_g^2 J_m \dot{\omega}_l + B_{eq} \omega_l = \eta_a \eta_m K_q K_t I_m$$
[3.7]

Finally, we can combine the electrical and mechanical equations by substituting [3.3] into [3.7], yielding our desired transfer function:

$$\frac{\omega_{l}(s)}{V_{m}(s)} = \frac{\eta_{g} \eta_{m} K_{t} K_{g}}{J_{eq} R_{m} s + B_{eq} R_{m} + \eta_{g} \eta_{m} K_{m} K_{t} K_{g}^{2}}$$
[3.8]

Where:

$$J_{eq} = J_l + \eta_g J_m K_g^2$$

This can be interpreted as the being the equivalent moment of inertia of the motor system as seen at the output.

Page # 4 Revision: 01

3.1 Pre-Lab Assignment

After deriving the system transfer function using mechanical first principals, it would be beneficial to understand the inherent control signals and accompanying states in the system.

Figure 2 below is a Simulink block diagram mapping the different stages of the system, each with their corresponding gains and conversions.

Using only the block diagram below, derive the transfer function from input (armature voltage) to output (load shaft velocity).

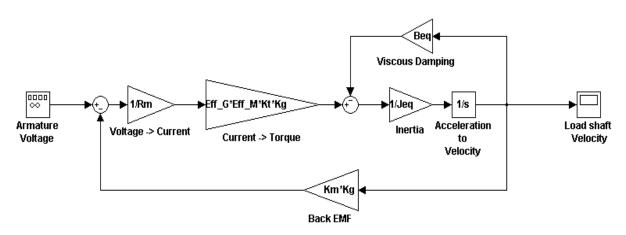


Figure 2 - Block Diagram of the SRV02 Plant

1) Derive the transfer function:

$$G(s) = \frac{\omega_i(s)}{V_i(s)}$$

2) Is the transfer function you derived using Figure 2 the same as the above transfer function of equation [3.8]?

Page # 5 Revision: 01

4. Lab Procedure

4.1 Wiring and Connections

The first task upon entering the lab is to ensure that the complete system is wired as described in the *SRV02 Experiment #0 - Introduction*. If you are unsure of the wiring, please refer to the *SRV02 Manual* or ask for assistance from a T.A. assigned to the lab. Now that all the signals are connected properly, start-up MATLAB and start Simulink. You are now ready to start the lab.

4.2 Controller Specifications

This lab requires you to design a Lead-Compensator to control the Servo plant with the following specifications:

- The system should have zero steady-state error (for a step input).
- The system Bandwidth should be 100 rd/s (approximately 16 Hz).
- The open-loop system should have a phase-margin of approximately 75 degrees.
- The system should have no over-shoot (very minimal at the most).

4.3 Open Loop Characteristics

When designing a controller of a system in the frequency domain, it is necessary to first study the open loop response of the system. In cases where the system cannot be modeled from first principles (or the system is too complicated), an open loop Bode plot is acquired by inputting a sinusoid of varying frequency and recording the **magnitude** and **phase** of the corresponding output. In this case, our model is of first order and our model is sufficiently accurate. The open-loop Bode plot to be used in the design will therefore be generated by MATLAB using our plant model.

In order to achieve a zero steady-state error to a step input, our system must be of type 1. By definition, a system is type 1 when it has a single pole located at the origin. We must therefore introduce an integrator in the loop to track a step input with no steady-state error.

We now have the transfer function G(s) with an integrator, yielding an open-loop transfer function of: G(s)

s

4.4 Controller Parameters

After achieved the first criteria of zero steady-state error (with the integrator), the next task should be to set the bandwidth of the system.

*Note: All relevant motor parameters can be found in Appendix A at the back of this handout or in the SRV02 User Manual.

Page # 6 Revision: 01

1) Generate a Bode plot of your open loop system and determine how much gain (Kp) must be introduced to achieve your desired system bandwidth.

*Hint: Your bandwidth will be the zero crossing of the open-loop Bode plot. Also, you should remember to use G(s) when generating your Bode plot.

s After you select a desired Kp, your new open-loop system is: $\underline{K_pG(s)}$

s

- Using your new open-loop system, generate another Bode plot to determine your phase-margin.
- Calculate the amount of phase your compensator would require to add such that the design specification is met.
- 2) Once you have decided on your parameters, Design a Lead compensator that adds the required phase at the required frequency. Refer to your in class material on how to go about designing such a compensator.

After determining your controller C(s), generate a Bode plot of your complete open loop system: $K_pC(s)G(s)$ to determine if all design specifications have been met.

Bode Diagram Gm = Inf, Pm = 75.456 deg (at 101.28 rad/sec) 40 20 Magnitude (dB) -20 -40 -60 -80 -45 -90 Phase (deg) -135 -180100 10² 103 10 10 Frequency (rad/sec)

Figure 3 - Open loop Bode plot of the complete system including the compensator. As you can see, the system has achieved a phase-margin of \sim 75 degrees as well as a bandwidth of \sim 100 rad/s

Page # 7 Revision: 01

Does your system meet the design specifications as seen in **Figure 3**? If you have not yet met the specifications, you should go back and redesign your compensator. Most control designs are iterative and will require fine-tuning to achieve the best control.

5. Simulation & Implementation

5.1 Controller Simulation

If you are content with the controller and have met the required design specifications, you are then ready to simulate your controller.

You must first run an M-file called "Setup_SRV02_Exp2.m". This file initializes all the motor parameters and gear ratios to the MATLAB workspace.

In Simulink, open a model called "s_tacho_lead.mdl". This model includes the plant model SRV-02 Plant Model (same as **Figure 2**), as a block in the final closed loop form. Use this model to simulate your controller.

You can implement your controller by making sure the following variables are set in the MATLAB workspace:

- Desired_Kp This variable should be set to the gain you had calculated to attain your desired bandwidth (section 4.4).
- Controller_NUM This is the numerator of your compensator in decreasing orders of s.
- Controller_DEN This is the compensator denominator in decreasing orders of s.

Ex: If your controller transfer function was:
$$C(s) = \frac{2s^2 + s - 1}{s^2 - 3s + 5}$$

Controller_NUM =
$$[2 \ 1 \ -1]$$

Controller_DEN = $[1 \ -3 \ 5]$

After setting the control parameters, you are now ready to simulate your design. Under the simulation menu, choose start. Monitor the *Simulated Velocity* scope and compare the signal to the *Setpoint Input*.

Looking at the system response (as well as the final Bode plot in section **4.4**), you should now have met all the design requirements. If you are content with your controller performance, and are satisfied with your simulation, you are now ready to implement your controller. Remember, if the simulated response is not desirable, you can easily re-iterate your design process until you reach a satisfactory simulation.

Page # 8 Revision: 01

5.2 Implementation of the Controller

Open a Simulink model called "q_tacho_lead.mdl". In this model, you will see 2 matching loops. This will allow you to run the simulation as well as the actual plant simultaneously, enabling you to note the difference in actual plant performance as opposed to the simulated plant.

You must now build the model using the WinCon menu. Once it has compiled, you may Start it using the WinCon server. The SRV-02 should now be tracking the command input. Plot the *Measured Velocity* as well as the *Setpoint* and the *Simulated Velocity*. This is done by clicking on the scope button in WinCon and choosing *Measured Plant Velocity*. Now you must choose the *Setpoint* and the *Simulated Velocity* signals thru the *Scope->File->Variables* menu.

You should now be monitoring all 3 signals on the same plot. If your controller had been designed to the above specifications and procedure, you should be seeing a response similar to **Figure 4** below.

*Note: Try adjusting the command input gain and take note of the response. **DO NOT** exceed a command of 60 RPM for the high gear configuration or a command of 300 RPM in the low gear configuration.

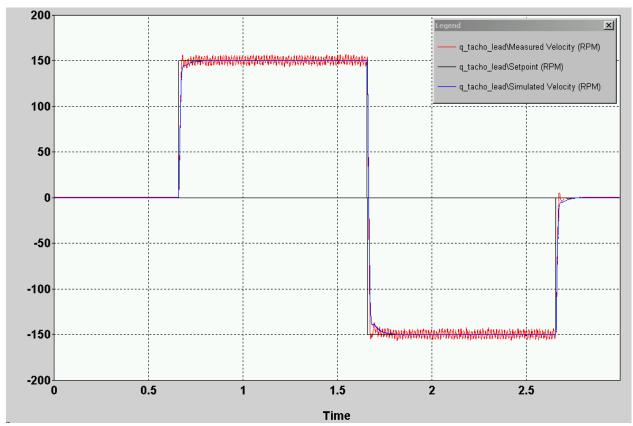


Figure 4 - Step response of the actual plant velocity (red) and the simulated velocity (blue) to an input command (black).

Page # 9 Revision: 01

6. Post Lab Questions & Report

After successfully designing and implementing your lead compensator, you should now begin to document your report. This report should include:

- I. System characteristic Bode plots. Here you should include all Bode plots you generated with all plots clearly labeled and documented.
- II. Design procedure. Clearly indicate all the steps you took in obtaining the final controller. In this section, make clear reference to which Bode plots you used and clearly label the system characteristic on each plot (i.e. phase margin).
- III. Any iterations in designing the controller should also be documented. (If your original design did not meet the required specifications or if you wanted a better response).
- IV. The following plots as they are instrumental in measuring the quality of your design. Make sure you include these plots and that they are labeled and documented:
 - A final open loop Bode plot of your complete system like the one seen in **Figure 3** of section **4**.
 - A step response to your system including the actual and simulated velocities as well as the input command. This should look similar to the **Figure 4** of section **5**.
- V. Final remarks and conclusions as well as answers to the post lab questions.

6.1 Post Lab Questions

- 1) Were there any design limitations or problems you encountered during the lab? If so, how were these limitations overcome?
- 2) Did the actual response of the system match directly to that of the simulated model? From your understanding of the system, what could you theorize is the source of these discrepancies and how would you improve them?
- If you were given the same speed control experiment again, would you choose a lead compensator as a controller or would you use a different approach? Explain.

Page # 10 Revision: 01

Appendix - A: SRV02 Nomenclature

Symbol	Description	MATLAB Variable	Nominal Value SI Units
$V_{\rm m}$	Armature circuit input voltage		
I_m	Armature circuit current		
$R_{\rm m}$	Armature resistance	Rm	2.6
$L_{\rm m}$	Armature inductance		
$\mathbf{E}_{ ext{emf}}$	Motor back-emf voltage		
θ_{m}	Motor shaft position		
ω_m	Motor shaft angular velocity		
$\boldsymbol{\theta}_{t}$	Load shaft position		
ω_l	Load shaft angular velocity		
$\mathbf{T}_{\mathbf{m}}$	Torque generated by the motor		
\mathbf{T}_{1}	Torque applied at the load		
K _m	Back-emf constant	Km	0.00767
\mathbf{K}_{t}	Motor-torque constant	Kt	0.00767
J_{m}	Motor moment of inertia	Jmotor	3.87 e-7
\mathbf{J}_{eq}	Equivalent moment of inertia at the load	Jeq	9.31 e-4
\mathbf{B}_{eq}	Equivalent viscous damping coefficient	Beq	1.5 e-3
\mathbf{K}_{g}	SRV02 system gear ratio (motor->load)	Kg	14 (14x1)
η_g	Gearbox efficiency	Eff_G	0.9
η_{m}	Motor efficiency	Eff_M	0.69
ω_n	Undamped natural frequency	Wn	
ζ	Damping ratio	zeta	
\mathbf{K}_{p}	Proportional gain	Kp	
\mathbf{K}_{v}	Velocity gain	Kv	
T_{p}	Time to peak	Тр	

Revision: 01